

GaN Ring Oscillators Operational at 500 °C Based on a GaN-on-Si Platform

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Abstract—A study of GaN for high temperature (HT, up to 500 °C) digital circuits was conducted. A HT-robust GaN-on-Si technology based on enhancement-mode p-GaN-gate AlGaIn/GaN high electron mobility transistors (HEMTs) and depletion-mode AlGaIn/GaN HEMTs was proposed and used to implement different digital circuit configurations, namely E/D-mode and E/E-mode (E: enhancement, D: depletion). The E/D-mode inverter was found to offer significantly better performance in terms of voltage swing, noise margin, and gain, across temperature and V_{DD} scaling. As calculated from E/D-mode ring oscillators (ROs) with $L_G = 2 \mu\text{m}$, a RO exhibited a propagation delay (t_p) of $< 1.48 \text{ ns/stage}$ at 500 °C. The best RO achieved $t_p < 0.18 \text{ ns/stage}$ at 25 °C. To the best of the authors' knowledge, the proposed technology sets a new boundary of t_p vs. L_G in wide band gap digital logic, and is operational at the highest reported temperature (500 °C) of a GaN digital circuit. The results reflect the promising potential of the proposed technology for emerging HT applications at 500 °C and beyond.

Index Terms—GaN, p-GaN-gate, transistor, high temperature, E/D-mode, E/E-mode, ring oscillator, propagation delay.

I. INTRODUCTION

EMERGING applications such as deep well oil drilling, hypersonic aircrafts, and exploration of Venus require high temperature (HT)-rated electronics components beyond the Si technology's typical temperature limit of 250 °C [1], [2]. Wide band gap semiconductors (SiC and GaN) are well suited to meet this demand thanks to their wide band gap and negligible carrier thermal generation at these temperatures [3]. While SiC HT digital circuits based on several transistor types have been proposed [4], [5], [6], [7], GaN and

III-N materials offer significant advantages in a wider range of applications from power [8], [9], [10] and RF [11], [12], [13], to MEMS [14], [15] and digital circuits [16], [17] across a large range of temperatures (from deep cryogenic temperature of 4 K to HT $> 1000 \text{ °C}$). In spite of these opportunities, their use in HT digital and analog circuits remains a relatively unexplored area.

For GaN digital circuits, an Enhancement-mode (E-mode) transistor is highly desired to avoid the need of an additional negative voltage supply (V_{SS}) [18]. This E-mode transistor could be realized using a number of technology options, including, (1) F-plasma treatment of gate region [19]; (2) recessed MIS gate [20]; (3) FinFET [21]; (4) p-GaN-gate [22]. Among these, the p-GaN-gate is especially interesting for HT robust operation thanks to the lack of gate dielectric (which degrades at HT and may introduce traps at the dielectric/semiconductor interface), and simplicity in process flow. Furthermore, at the technology platform level, the p-GaN-gate AlGaIn/GaN HEMT offers the possibility of monolithic integration with both depletion-mode (D-mode) n-FETs ([23], [24]) and E-mode p-FETs ([16], [17], [25], [26]).

In this work, HT digital circuits were realized based on p-GaN-gate AlGaIn/GaN HEMTs which are optimized for HT operation and have been demonstrated to offer robust performance at least up to 500 °C [27]. The DC (static) performance of inverters with two different circuit configurations (E/D-mode and E/E-mode) was studied. Based on the results at the inverter-level, ring oscillators (ROs) were demonstrated in order to understand the propagation delay (t_p), an important performance metric for HT digital circuits.

II. CHOICE OF INVERTER CIRCUIT CONFIGURATION

As illustrated in Fig. 1(a), the wafer platform used in this work is p-GaN/AlGaIn/GaN-on-Si and allows for the monolithic integration of two types of transistors, E-mode p-GaN-gate AlGaIn/GaN HEMTs and D-mode AlGaIn/GaN HEMTs, using the process flow described in [28]. Here, HT transistors with refractory metal gate and self-alignment in p-GaN-gate were fabricated. The typical transfer characteristics of the E-mode and D-mode transistors ($L_G = L_{GS} = L_{GD} = 2 \mu\text{m}$) are shown in Fig. 1(b). Good ON-OFF ratio ($> 3 \times 10^7$, limited by gate leakage) and V_{th} of 1.4 V at room temperature are obtained for the E-mode transistors. The D-mode transistor shows a V_{th} of -1 V .

In order to identify the optimal implementation of the GaN high temperature logic, the characteristics of two classic inverter configurations (E/D-mode and E/E-mode inverters) (Fig. 1(c)) were experimentally evaluated at high temperature.

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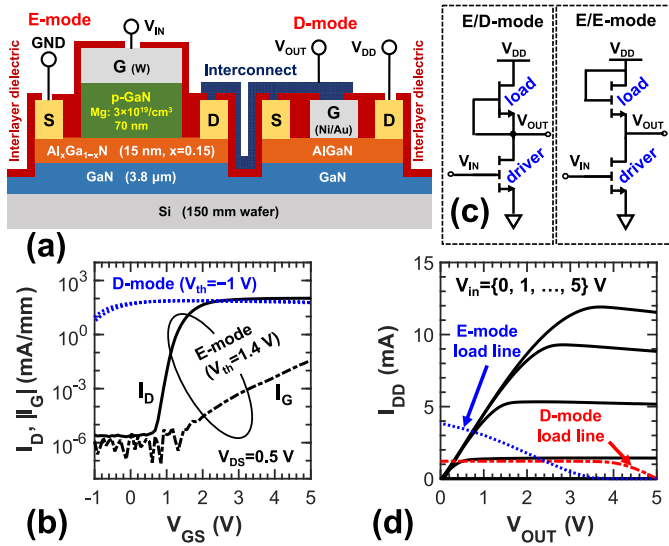


Fig. 1. Transistor technology and circuit configurations of *n*-FET only logic. (a) Illustration of the E-mode transistor (p-GaN-gate AlGaIn/GaN HEMT) and D-mode transistor (AlGaIn/GaN HEMT) connected as an E/D-mode inverter. In all transistors, $L_G = L_{GS} = L_{GD} = 2 \mu\text{m}$. (b) Typical transfer characteristics of the E-mode and D-mode transistors. (c) Circuit configurations of E/D-mode and E/E-mode logic. (d) IV curves of an E-mode transistor ($W/L = 36/2 \mu\text{m}/\mu\text{m}$). $I_{D,max} = 330 \text{ mA/mm}$, $R_{ON} = 12 \Omega\cdot\text{mm}$ (measured at $V_{GS} = 5 \text{ V}$). The load line, realized using a either D-mode or E-mode transistor ($W/L = 12/2 \mu\text{m}/\mu\text{m}$), is also included. The above measurements are conducted at room temperature.

In principle, the E/E-mode inverter offers: (1) higher simplicity, (2) higher current, and (3) higher speed at higher V_{DD} , while the E/D inverter features (1) lower power consumption, (2) better gain, voltage swing (V_{swing}), and noise margin (NM), and (3) higher speed at lower V_{DD} , based on basic digital circuit theory and the prior work in other semiconductor technologies [29]. In the inverters of this work, the driver and load transistors are set to $W/L = 36/2 \mu\text{m}/\mu\text{m}$ and $12/2 \mu\text{m}/\mu\text{m}$, respectively. The transistor sizing is determined based on: (1) the characteristics of the load line (Fig. 1(d)), (2) trade-offs between symmetry in the DC (static) and transient (dynamic) characteristics of the inverter, (3) tolerance to non-uniformity and temperature variation in transistor characteristics, and (4) layout area. For a fair comparison between E/D-mode and E/E-mode, the same sizing is used.

A comparison of the voltage transfer curves (VTCs, Fig. 2(a)) at 300 °C reveals that, the E/D-mode inverter features significantly better performance than the E/E-mode inverter, in terms of V_{swing} ($= V_{OH} - V_{OL}$), gain, and NM . This is because: (1) In E/D-mode, $V_{OH} = V_{DD}$, whereas in E/E-mode, $V_{OH} = V_{DD} - V_{th(E)}$; (2) In both circuit configurations, V_{OL} is limited by the voltage drop across the E-mode driver. While the R_{ON} of both drivers is similar, the current drive of the E-mode load (diode-connected), when input is HIGH, is significantly stronger than that of the D-mode load (GS-tied) (Fig. 1(d)).

The above observations are also valid with the scaling of V_{DD} , an important design parameter in digital circuits (Fig. 2(b)–(c)). With the increase in V_{DD} , the V_{swing} of E/D-mode and E/E-mode inverters generally improves. However, the magnitude of improvement is better for the E/D-mode, thanks to the constant V_{OL} across different V_{DD} .

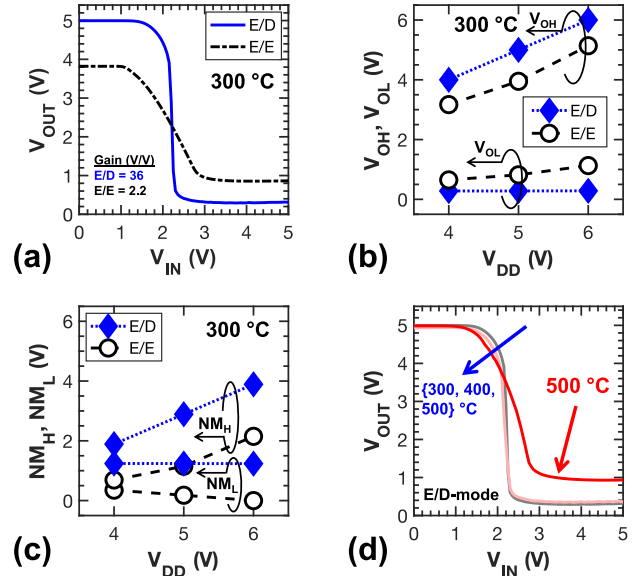


Fig. 2. Comparison of E/D-mode and E/E-mode inverters. (a) VTC at 300 °C. Maximum voltage gains of E/D-mode and E/E-mode are 36 V/V (at $V_{in} = 2.1 \text{ V}$) and 2.2 V/V (at $V_{in} = 2.6 \text{ V}$), respectively. (b) V_{OH} , V_{OL} vs. V_{DD} for E/D-mode and E/E-mode at 300 °C. V_{swing} may be calculated by $V_{OH} - V_{OL}$. (c) Noise margins (NM_L , NM_H) vs. V_{DD} for E/D-mode and E/E-mode at 300 °C. (d) VTC of E/D-mode up to 500 °C.

When the output is LOW, the R_{ON} of the E-mode driver is almost independent of V_{DD} (Fig. 1(d)). To first order, the D-mode load supplies a constant current, compared to the current of E-mode load with V_{DD}^2 dependency, leading to a better V_{OL} at higher V_{DD} .

Having established that E/D-mode technology offers significantly better performance in GaN digital circuits at 300 °C, the temperature dependence of E/D-mode was further studied for higher temperatures. Measurements beyond 300 °C were conducted using a probe station with a hot chuck (maximum rating of 500 °C) and sealed chamber in N_2 ambient. As presented in Fig. 2(d), the VTC of the E/D-mode remains largely constant up till 400 °C, though NM_L is reduced slightly (by $\sim 0.15 \text{ V}$) due to the faster degradation in the ON-resistance of the E-mode driver transistor than in the D-mode load transistor. At 500 °C, an increase in V_{OL} (hence reduction of V_{swing} by 0.6 V_{pp}) was observed, which can also be explained by the faster degradation of the E-mode device ON-resistance.

III. HIGH TEMPERATURE RING OSCILLATOR AND PROPAGATION DELAY

In order to estimate the t_p of the proposed GaN E/D-mode technology, ROs were fabricated with $(W/L)_{\{E,D\}} = \{36/2, 12/2\} \mu\text{m}/\mu\text{m}$ (Fig. 3(a)). t_p was calculated by $T_{RO}/2N_{RO}$, where T_{RO} is the period of the RO waveform V_{OUT} , N_{RO} is the number of stages of the RO. The V_{DD} scaling trends of ROs were compared at 300 °C (Fig. 3(b)). The t_p may be modeled as $\frac{1}{2}C_L \times V_{DD}/I_{ave}$, where C_L is the load capacitance, and I_{ave} is the average charge/discharge current. To first order, the charge current through the D-mode load is constant with V_{DD} (refer to Fig. 1(d)), therefore t_p from LOW to HIGH (t_{pLH}) is proportional to V_{DD} . The discharge current through the E-mode driver, same as Si CMOS circuits, is roughly proportional to V_{DD}^2 , therefore $t_{pHL} \propto 1/V_{DD}$.

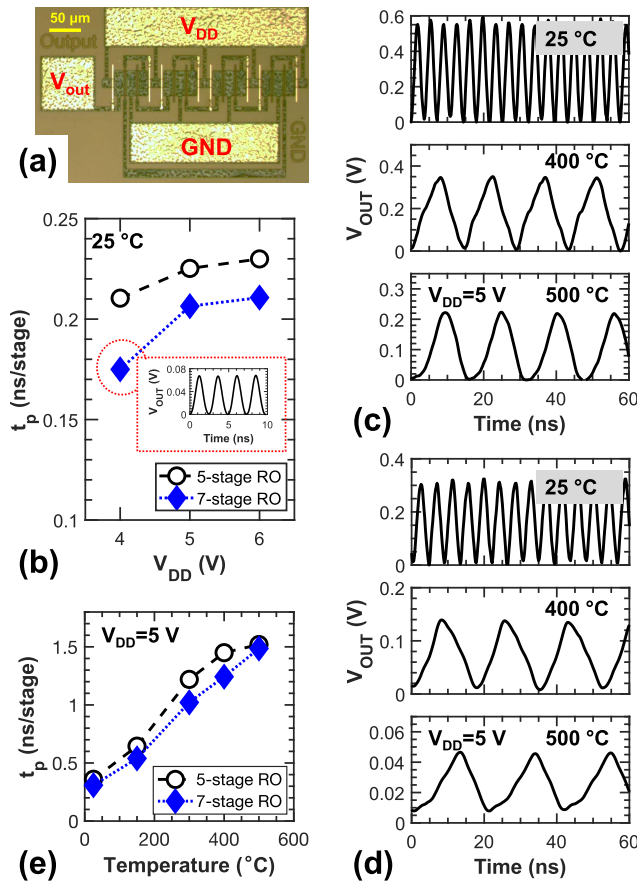


Fig. 3. GaN ring oscillators (ROs). (a) Micrograph of a 7-stage RO. (b) V_{DD} scaling trend of the best 5-stage RO and 7-stage RO at 25 °C. The inset presents the output waveform V_{OUT} of the 7-stage RO which demonstrates the best $t_p < 0.18$ ns/stage in this work. (c) Waveforms of another 5-stage RO at various temperatures. (d) Waveforms of another 7-stage RO at various temperatures. (e) t_p vs. temperature, estimated from the data reported in (c)–(d).

Due to the significantly lower charge current, t_{pLH} dominates t_p . Hence, a lower V_{DD} results in a lower t_p . However, the reduction of V_{DD} results in trade-offs in NM and V_{swing} (Fig. 2(c)–(d)). Therefore, V_{DD} of 5 V was chosen for the RO as a compromise.

The ROs were operational at 500 °C, as shown in Fig. 3(c). The degradation of t_p from 0.31 ns/stage (25 °C) to 1.48 ns/stage (500 °C) (Fig. 3(d)) may be attributed to the performance degradation of both E/D-mode transistors (increase in R_{ON} of the E-mode driver, reduction in current drive capability of the D-mode load) and an increase in parasitics.

The best RO of this work ($t_p < 0.18$ ns/stage, Fig. 3(b) inset) was benchmarked against other ROs based on wide band gap electronics (GaN [17], [19], [22], [30], [31], [32], [33] and SiC [4], [5], [6], [7]). As shown in Fig. 4(a), to the best of the authors' knowledge, at room temperature, the proposed technology sets a new boundary in the well-known relationship of t_p vs. L_G^2 [34]. Furthermore, as shown in Fig. 4(b), the reported RO is operational at the highest reported temperature (500 °C) of a GaN digital circuit. The results reflect the promising potential of the proposed technology, which is based on p-GaN-gate AlGaIn/GaN HEMTs optimized for HT (≥ 500 °C) applications.

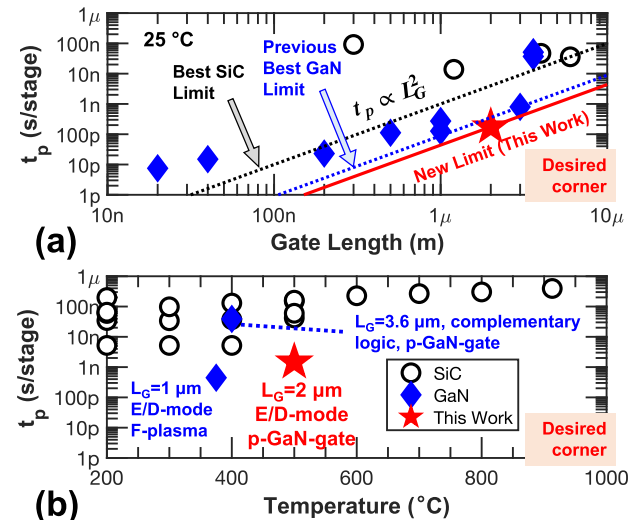


Fig. 4. A summary of ROs reported in the literature based on wide band gap electronics (GaN [17], [19], [22], [30], [31], [32], [33] and SiC [4], [5], [6], [7]). (a) t_p vs. L_G . The general scaling trends ($t_p \propto L_G^2$) for the best SiC demonstration, previous best GaN demonstration and the best result of this work (Fig. 3(b) inset) are included for reference. The L_G of the pull-down transistor is chosen. (b) t_p vs. temperature. The data points showing GaN RO demonstrations are labelled with the L_G of the pull-down transistor, the logic family/circuit configuration and the type of E-mode n-FET. To the best of the authors' knowledge, the proposed technology in this work defines a new boundary of t_p vs. L_G , as well as the operating temperature of GaN digital circuits.

The measured values of t_p should be considered as an upper limit for the intrinsic t_p . As verified in Fig. 3(d), ROs with more stages would give a more accurate estimation of the intrinsic t_p . In the RO, in addition to the odd number of inverters connected in a circular chain, an output buffer is used for the readout of the output waveform (V_{OUT}). This buffer stage introduces a fixed delay whose relative contribution decreases as the number of stages increases. In this experiment, the number of stages in the E/D-mode RO was limited by the uniformity and yield of the fabrication. On the other hand, for E/E-mode ROs, the number of stages is limited by the low gain and low noise margins (especially NM_L) of each E/E-mode inverter stage (Fig. 2(a)), which makes the oscillations extremely difficult at high temperature.

IV. CONCLUSION

A study of E/D-mode and E/E-mode inverters realized by HT-optimized E-mode p-GaN-gate HEMT technology based on a GaN-on-Si platform was conducted. E/D-mode inverters were found to offer significantly higher performance than E/E-mode inverters at 300 °C. The reported RO exhibited a t_p of < 1.48 ns/stage at 500 °C. The best RO achieved $t_p < 0.18$ ns/stage at 25 °C with $L_G = 2$ μm . Further advancement of the proposed technology, e.g. optimization of the epitaxial structure, reduction of layout parasitics, and introduction of HT-rated advanced packaging [35], could significantly push the performance limit of GaN HT electronics.

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